

15W Wireless Power Receiver With 5W Transmitter

Description

The CV8045D is a highly-integrated Wireless power single-chip (SoC) with 15W receiver/5W transmitter dual modes, The device can be configured to receiver or transmitter modes, when the devices work in transmitter mode, The rectifier bridge works as a full/half bridge inverts, has a 16bits PWM generator, with dead-zone regulating. Embedded dual channels demodulation for Communication. When device work as receiver mode, The device receives an AC power from a wireless transmitter, embedded a high efficiency synchronous full bridge rectifier converts to DC power, Embedded modulation and FSK demodulation circuits supports bi-direction communication,

A Microprocessor manages power receiver/transfer and communication, through a I2C port connect with Mobile phones AP, Embedded 16K Bytes MTP memory to support on line debugging and OTA function, low power design to meet the ENERGY star Requirements.

The device includes over-temperature, over-current and over/under voltage protections. The CV8045D is available in 53-WLCSP package with a thickness of only 0.3mm.

Wireless power system

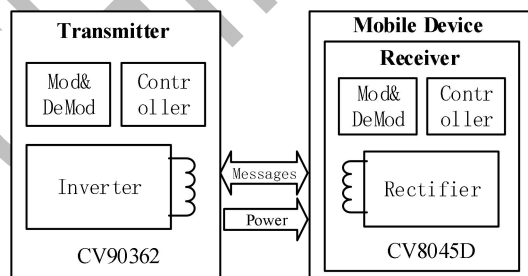


Figure 1

Features

- Single-chip with Rx/TX dual modes wireless power solution
- Compliance with WPC V1.2.4 BPP&EPP
- 16K bytes Multiple-time programmable(MTP) no-volatile memory
- Support private profile/protocol extension
- Support I2C 400KHZ standards interface
- Embedded 12bits high accuracy ADC
- Embedded $\pm 1.5\%$ accuracy OSC in full temperature rank
- Low standby and operating power consumption
- 8 GPIOs with 1 channel Capture
- 6 channels external ADCs
- Vrect input voltage up to 20V
- Maximum continuous operating current 1.5A
- Receiver mode
 - Delivers up to 15W as a receiver
 - High Efficiency synchronous rectifier with low Rds(on)
 - Low dropout regulator with low Rds(on)
 - High accuracy frequency detection for FSK communication
 - Output Voltage up to 20V, with 20mV regulation/step from 3.6V~20V
 - Programmable current limit
 - Programmable Clamping voltage and strength
- Transmitter mode
 - Up to 5W power delivery
 - Invert bridge with Half/Full bridge control mode
 - High accuracy PWM controller to support phase-shift, PWM modulation and dead-zone control
- Package: 6x9 Ball array, 2.7mm x 4.0mm x 0.3mm, 53-WLCSP with 0.4mm ball pitch,

Typical Applications

- Wireless power RTx solution for portable devices
- Mobile phone
- Power bank
- Tablets
- Accessories
- Stationary device power supply

1 System diagram

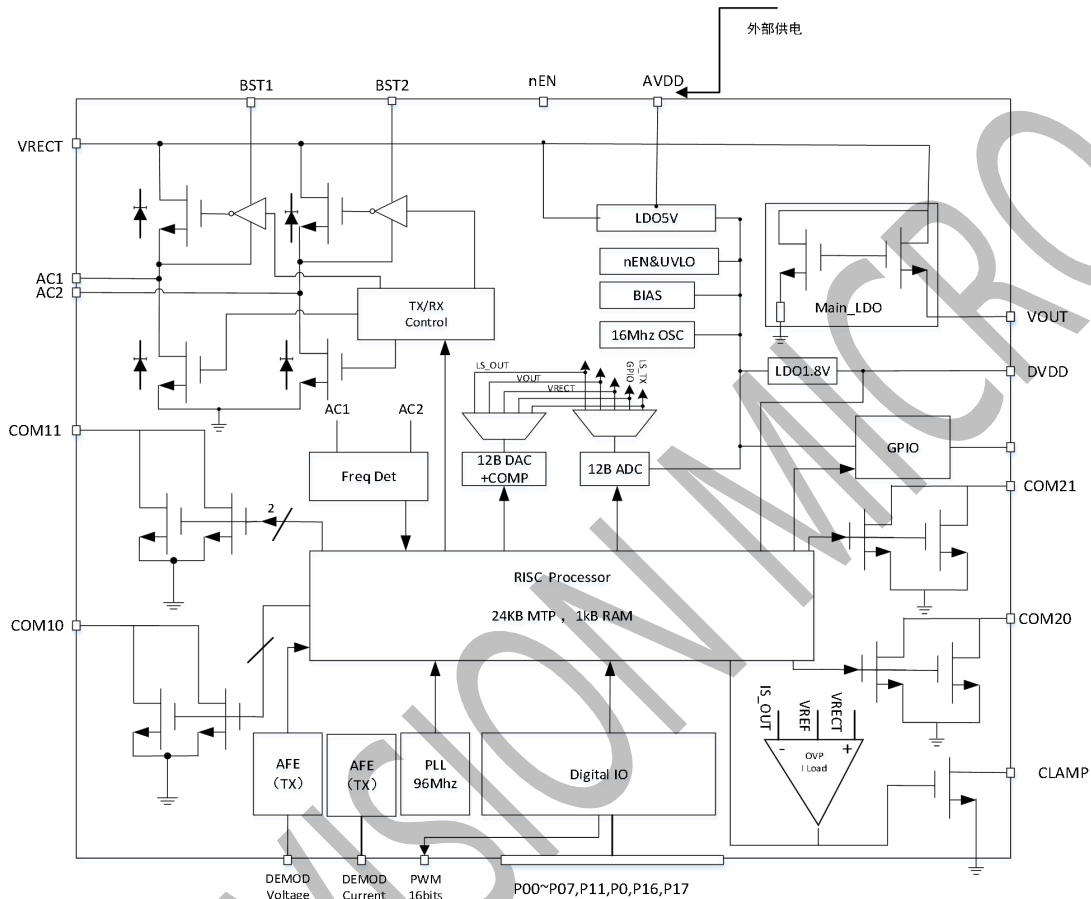


Figure 2

Wireless power system description

The CV8045D is a highly-integrated wireless power transmitter and receiver dual-function IC for mobile or stationary devices, The internal block diagram of CV8045D is shown in figure 2.

The device can transmit up to 5W in WPC Transmitter mode and receiver up to 15W power, It also can support PMA mode (option).

Rx mode

In Rx mode, The device receive the wireless power and stored on Vrect capacitors, until the voltage exceeds the UVLO threshold, the rectification is performed by the body diodes of Rectifier FETs, Once the internal biasing circuit is working, The CPU can enable high-side FETs, low-side FETs or full bridge ETs to work in switching mode, Depend on system status and power demand. The device uses 12bits ADC to monitor the Vrect and load current, To send instructions to wireless power

transmitter for increasing or decreasing power. The output voltage of low-drop-out regulator (Main LDO) is programmable from 3.6V to 20V with 20mV/step using I2C commands. The OVP, OTP, OCP protection circuit are continuously monitor to ensure system work well.

Tx mode

In Tx mode, The power will input through Vout pin, The processor will control Rectify full bridge to work as inverter bridge, A high resolution PWM can work as 128mHZ to control the inverter bridge work as half/full bridge modes, and dead-zone setting, A high resolution Voltage decoding and A high resolution current decoding to get Rx's instruction through DEMOD filter shown in Figure 3, Based on the receive packet, the CV8045D will adjust the operating frequency to match the transmitted power level to reliable wireless power transfer, A high resolution PWM controller ensure system in higher efficiency.

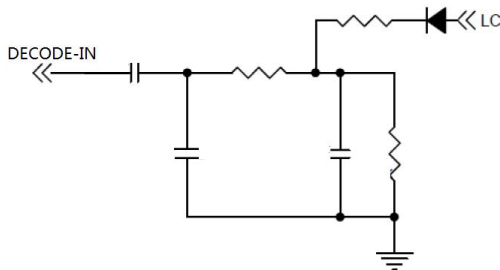


Figure 3

Power Control

The voltage across the Vrect and the current through the rectifier are periodically sampled and digitized by the ADC. The digital equivalent of voltage and current is provided to internal control logic that determines if the operating point needs to be changed based on the load conditions on the Vrect. If the load is heavy enough to cause the voltage at Vrect to be below the target value, the power transmitter is instructed to reduce its frequency, near resonance. If the voltage at Vrect is above its target, the power transmitter is instructed to increase its frequency. To maximize efficiency, the voltage of the Vrect is programmed to decrease as the LDO load current increases.

Power Transfer

When a mobile device containing the CV8045D is placed on the WPC "Qi" charging pad, the CV8045D responds to the transmitter's "Ping" signal, completing the "Identification and Configuration". Once the "Identification and Configuration" phase is completed and successfully negotiated and calibrated, the transmitter will initiate the power transfer mode. The CV8045D control circuit measures the Vrect voltage and sends a control error packet to the power transmitter, adjusting the Vrect voltage to achieve optimum efficiency of the LDO. At the same time, the rectifier bridge power data packet is sent to the power transmitter as the basis of the transmitter foreign object detection (FOD) to ensure safe and effective power transmission.

Synchronous Rectifier

The CV8045D's built-in synchronous rectifier bridge can effectively improve the rectification efficiency. When the load is higher than 100mA, the rectifier bridge operates in a synchronous full-bridge rectification manner. When the load is less than 100 mA, the rectifier bridge operates in a semi-synchronous full-bridge rectification mode. During the power-on phase, when the V_{rect} voltage is lower than the uv_{lo} threshold, the rectifier bridge is fully bridge rectified by the NMOS body diode. The BST capacitor is used to provide a switching drive voltage for the NMOS of the upper bridge.

Advanced Foreign Object Detection (FOD)

When the metal is placed in an alternating magnetic field, the electromagnetic eddy current heats the metal. For example, coins, keys, paper clips, etc. The degree of heating depends on the amplitude and frequency of the coupled magnetic field, as well as the resistible, size and shape of the object. In wireless energy transmission systems, this heat is energy loss, reducing energy transfer efficiency. If proper measures are not taken, metal objects are continuously heated and high temperatures are generated, which may cause other dangerous situations.

In addition, there may be other metals in the final product design of the WPC power transmitter and receiver (these metals are neither part of the power transmitter nor part of the power receiver, but will be from the coupled AC magnetic field during power transmission). Absorbing energy, causing power loss, such as lithium-ion batteries, metal ICs, etc., so FOD detection also needs to compensate for the power loss caused by these metals.

The CV8045D uses advanced FOD technology to detect foreign objects placed on or near the launch pad. The FOD settings can be optimized through an I2C interface or programming to match the power transfer characteristics of each particular WPC system, including power losses for TX and RX coils, batteries, shields, and housing materials from no load to full load. These values are based on a comparison of the received power to the reference power curve so that any foreign matter can be detected when the received power is different than the expected system power.

Over-voltage Protection

If the input voltage increases above setting value (15V、22V、27V), the control loop disables the LDO, sends a control error packet to the power transmitter to attempt to restore the rectifier voltage to a safe operating voltage level, and uses high voltage open drain (Clamp/Sink) to control the OVP FET to the input voltage. Clamping allows V_{rect} to stabilize. The clamp is released when the V_{rect} voltage is below the $VOVP$ hysteresis calibration level.

Over temperature、 Over current Protection

Both the over temperature protection threshold and the over current protection threshold of the CV8045D can be programmed. When the output current of the CV8045D exceeds the over current

protection threshold or the detected temperature exceeds the over temperature protection threshold, the CV8045D turns off the LDO output and sends a charge end packet to the power transmitter to terminate the power transfer.

Status Output

GPIO2-4 can be selected to indicate the current working status. For example, charging is completed, charging is abnormal, and the like.

LDO

The CV8045D has three LDOs built in, a high-power LDO, programmable select outputs of 3.6V--20V, VDD5V LDO and VPP18 LDO (VDD5V and VPP18 are both used to power the internal low-voltage operating modules). A filter capacitor is required on each LDO pin.

WPC Mode Communication

Modulation method

According to the WPC specification, in the wireless medium power transmission system, the duplex communication method is adopted: the communication sent by the receiver to the transmitter - the amplitude shift keying (ASK) and the communication sent by the transmitter to the receiver - frequency shift keying (FSK) .

The communication signal sent by the receiver to the transmitter is controlled by the baud rate of 2kpbs to control the external capacitor connected to the internal switch and AC1/AC2 to be grounded or suspended to adjust the load on the receiving induct coil for modulation. This causes the output impedance of the transmitter to change, and this communication signal is ultimately reflected in the resonant amplitude of the transmitting coil. The transmitter acquires communication data by detecting changes in voltage or current on the transmitting coil.

The communication signal sent by the transmitter to the receiver is modulated by changing the frequency of the AC power signal of the transmitter. The receiver detects a frequency change and acquires communication data. A handshake protocol is established with the transmitter through this communication data.

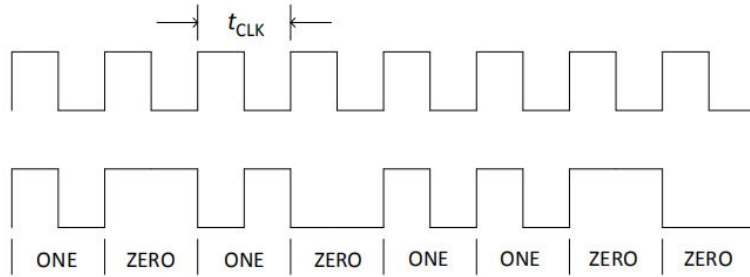
Data Format

According to the WPC specification, the CV8045D communicates with the power transmitter or receiver in the form of data packets. The format of the data packet is as follows:

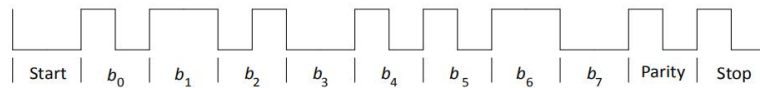
Preamble	Header	Message	Checksum
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Encoding

According to the WPC specification, the CV8045D uses a 2 kHz clock frequency to modulate data bits onto the power signal using a two-phase differential encoding. Logic one uses two narrow transforms for encoding, while logic zero uses two wide transforms for encoding, as follows:



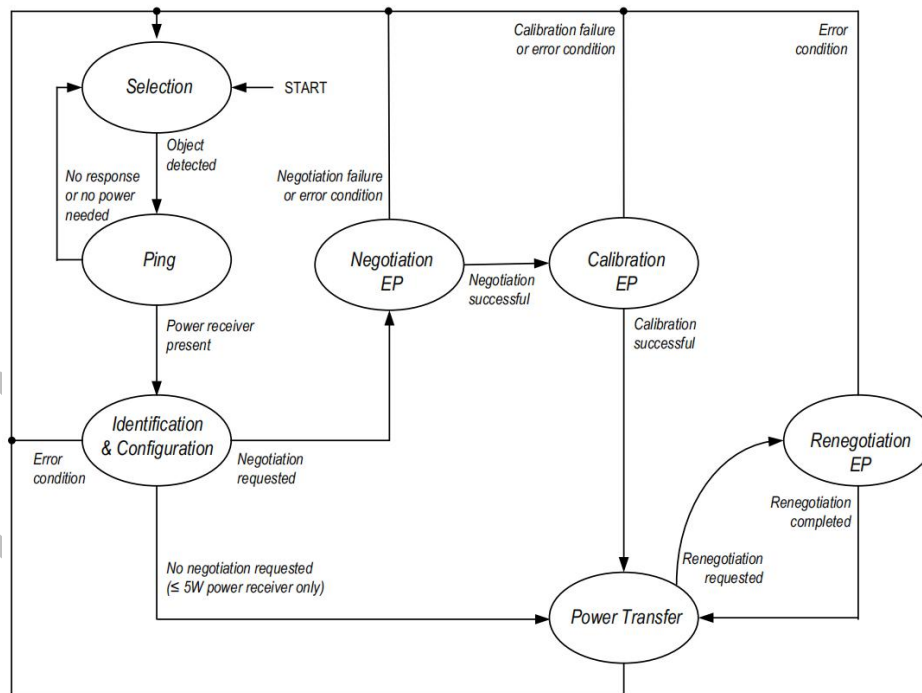
Each byte in the communication packet includes 11 bits in the asynchronous serial format as follows:



System Feedback Control

The CV8045D is fully compatible with WPC (latest specification) and has all the necessary circuitry to communicate with the transmitter or receiver via a WPC communication packet.

The communication process between the transmitter and the receiver is as follows:



The CV8045D goes through five phases:

Selection, Ping, Identification & Configuration, Negotiation, Calibration & Power Transfer

Selection

At this stage, the CV8045D receives or transmits wireless power and enters the ping phase. When the Vrect voltage is higher than UVLO, the CV8045D is ready to communicate with the transmitter or enter the power ping mode.

Ping

At this stage, the CV8045D sends a signal strength packet as the first communication packet to instruct the sender to keep the power signal on (or the CV8045D detects the signal strength packet). After transmitting/receiving the signal strength packet, the CV8045D enters the identification and configuration phase. Conversely, if a transport end packet is sent, it will remain in the ping phase. At this stage, the following two messages are sent/expected:

- Signal strength packet
- End of power packet

Identification & Configuration

At this stage, the following two messages are sent/expected:

- Identification packet
- Configuration packet

NEGOTIATION

The receiver negotiates with the transmitter to adjust the transmitter. In this process, the receiver sends a negotiation request to the transmitter, and the transmitter can agree or reject the negotiation request.

CALIBRATION

At this stage, the receiver provides the received power to the transmitter.

POWER TRANSFER

- At this stage, the CV8045D controls power transfer through the following control packets:
- Control error packet
- Rectified power packet
- End power transfer packet

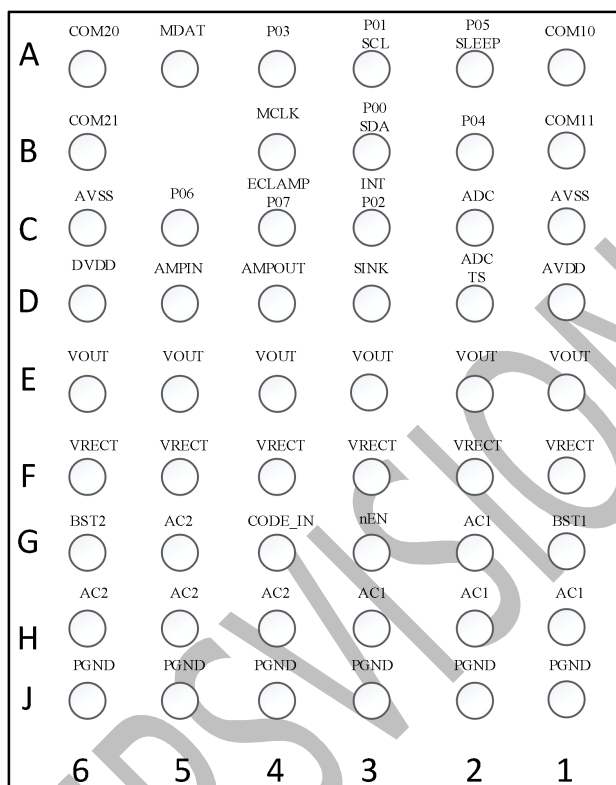
RE-NEGOTIATION

At this stage, the receiver can communicate with the transmitter to adjust if needed. This phase can be terminated early without changing the transmit power.

END OF POWER

When the load on the receiver ends the power request (eg, charging is completed), the CV8045D turns off the LDO output, and continuously transmits the transmission end packet to the transmitter until the transmitter ends the power transmission, or the receiver's Vrect voltage is lower than the UVLO threshold.

2 Pin assignments



Bottom View

3 Pin descriptions

CV8045D Pin Definition (Ball View)

Ball NO	Pin Name	I/O	Description
A1	COM10	O	Communication modulation signal output
A2	P05 SLEEP	I/O	GPIO, Can be set as a 16bits PWM output port,

Ball NO	Pin Name	I/O	Description
A3	P01 SCL	I/O	I2C clock pin. Open-drain output. Connect a 5.1k Ω resistor to VDD18 pin.
A4	P03	I/O	Open Drain GPIO
A5	MDAT	I/O	Program data
A6	COM20	O	Communication modulation signal output
B1	COM11	O	Communication modulation signal output (Reserved)
B2	P04	I/O	Open Drain GPIO
B3	P00 SDA	I/O	I2C data pin. Open-drain output. Connect a 5.1k Ω resistor to VDD18 pin.
B4	MCLK	I	Program CLK, connect a 10K Ω resistor to VDD18 pin.
B5	NC		
B6	COM21	O	Communication modulation signal output
C1	AVSS	A	Analog GND
C2	ADC P12	I/O	ADC
C3	P02 INT	I/O	Open Drain GPIO
C4	ECLAMP ADC P07	I/O	Push-Pull output driver for External Power Clamp FET gate control (Connect a resistor from Vrect to the external FET to GND). This pin can be floating if not used.
C5	P06	I/O	Open Drain GPIO, Can be set to capture
C6	AVSS	A	Analog GND
D1	AVDD	A	Internal 5V regulator output voltage. Connect a 1 μ F capacitor from this pin to ground.
D2	ADC(TS)	A	External ADC channel
D3	SINK (CLAMP1)	O	Open drain output for over voltage protection, Which will be trigger, once the Vrect over setting voltage. Connect a resistor from this pin to the Vrect pin, for more detail information about over voltage settings, see section 6.1
D4	AMPOUT	A	Communication signal for current decoding AMP output on TX_mode, can be floating once the

Ball NO	Pin Name	I/O	Description
D5	AMPIN	A	Current decoding input on TX_mode, connect to GND if the Current decoding channel has not be used.
D6	DVDD	A	Internal 1.8V regulator output voltage
E1	VOUT	A	Output voltage to load
E2	VOUT		
E3	VOUT		
E4	VOUT		
E5	VOUT		
E6	VOUT		
F1	VRECT	A	Output voltage of the synchronous rectifier bridge. Connect three 10 μ F capacitors from this pinto GND
F2	VRECT		
F3	VRECT		
F4	VRECT		
F5	VRECT		
F6	VRECT		
G1	BST1	A	Boost capacitor for driving the high-side switch of the internal rectifier, Connect a 15nFcapacitor from the AC2 pin to BST2
G2	AC1	A	AC input power. Connect to the resonant capacitor
G3	nEN	I	Chip enable pin
G4	DECODE-IN	I	Voltage decoding input on TX_mode
G5	AC2	A	AC input power. Connect to the Rx coil
G6	BST2	A	Boost capacitor for driving the high-side switch of the internal rectifier. Connect a 15nFcapacitor from the AC1 pin to BST1
H1	AC1	A	AC input power. Connect to the resonant capacitor
H2	AC1		
H3	AC1		
H4	AC2	A	AC input power. Connect to the resonant capacitor
H5	AC2		

H6	AC2		
J1	PGND	P	GND
J2	PGND		
J3	PGND		
J4	PGND		
J5	PGND		
J6	PGND		

4 Electrical specification

4.1 Absolute Maximum Ratings

Symbol/Pins	Parameter	Minimum	Maximum	Units
COM21, COM20, COM11, COM10, SINK (CLAMP1), VRECT, AC1, AC2	Maximum voltage	-0.3	36	V
BST1, BST2	Maximum voltage	-0.3	AC1+5, AC2+5	V
DVDD, P05 (SLEEP), P04, MDAT, P02/INT, P06, ADC(TS)	Maximum voltage	-0.3	2	V
AVDD, DECODE-IN, nEN, AMPOUT, SCL, MCLK, SDA, P03	Maximum voltage	-0.3	6	V
PGND	Maximum voltage	-0.3	0.3	V
COM21, COM20, COM11, COM10	Maximum RMS current		500	mA
AC1, AC2	Maximum RMS current		3.6	A
VOUT Output Current	Maximum RMS current		3.6	A
HBM	ESD – Human Body Model		2000	V
CDM	ESD – Charged Device Mode		500	V

Absolute maximum ratings are those values beyond which damage to the device may occur. Functional operation under these conditions is not implied.

4.2 Thermal Characteristics

Symbol	Parameter	Value	Units
θ_{JA}	Thermal Resistance Junction to Ambient	45	°C/W
θ_{JC}	Thermal Resistance Junction to Case	0.2	°C/W
θ_{JB}	Thermal Resistance Junction to Board	4.36	°C/W
T_J	Operating Junction Temperature	-5 to +150	°C
T_{AMB}	Ambient Operating Temperature	0 to +85	°C
T_{STOR}	Storage Temperature	-55 to +150	°C
T_{BUMP}	Maximum Soldering Temperature (Reflow, Pb-Free)	260	°C

[a] The maximum power dissipation is $PD(MAX) = (T_J(MAX) - T_A) / \theta_{JA}$ where $T_J(MAX)$ is 150°C. Exceeding the maximum allowable power dissipation will result in excessive die temperature, and the device will enter thermal shutdown.

[b] This thermal rating was calculated on JEDEC 51 standard 4-layer board with dimensions 3" x 4.5" in still air conditions.

[c] Actual thermal resistance is affected by PCB size, solder joint quality, layer count, copper thickness, air flow, altitude, and other unlisted variables.

4.3 Electrical Characteristics Table

Symbol	Description	Conditions	Min	Typ	Max	Units
Input Supplies & UVLO (Tx Mode)						
V_{IN_OUT}	V_{OUT} Input Operating Voltage Range		3.6	5	20	V
V_{IN_UVLO}	Under-Voltage Lockout	VIN Rising	2.6	2.8	3.0	V
		VIN Failing		200		mV
I_{SHD}	Shutdown Current	$V_{EN} = V_{IN}$		500		uA
Input Current Sense (Tx Mode)						
V_{SEN_OFST}	Amplifier Output Offset voltage	Measured at amplifier output node; $V_{ISH} = V_{ISL}$		0.6		mV
I_{SENACC_TYP}	Measured Current sense accuracy	$V_{R_ISNS} = 10mV$		±2		%
Analog to Digital Converter						
N	Resolution			12		Bit

f_{sample}	Sampling Rate			67.5		kSa/s
Channel	Number of channels			12		
$V_{\text{IN,FS}}$	Full scale Input voltage			5		V
Thermal Shutdown						
TSD	Thermal shutdown	Threshold Rising		140		°C
		Threshold Falling		120		°C
Clocks						
F_{LSOSC}	System clock			16		Mhz
General Purpose Inputs/Outputs						
V_{IH}	Input threshold high		1.35			V
V_{IL}	Input threshold low			0.5		V
I_{LKG}	Input Leakage Current	0V and 1.8V	-1		3	uA
V_{OH}	Output logic high	$I_{\text{OH}} = 4\text{mA}, 12\text{mA total}$	1.44			V
V_{OL}	Output logic low	$I_{\text{OL}} = 12\text{mA}$			0.36	V
SCL, SDA (I ² C Interface)						
f_{SCL}	Clock Frequency				400	khz
$t_{\text{HD,STA}}$	Hold Time (Repeated) for START Condition		0.6			us
$t_{\text{HD,DAT}}$	Data Hold Time		0			ns
t_{LOW}	Clock Low Period		1.3			us
t_{HIGH}	Clock High Period		0.6			us
$t_{\text{SU,STA}}$	Set-up Time for Repeated START Condition		0.6			us
t_{BUF}	Bus Free Time Between STOP and START Condition		1.3			us
C_{B}	Capacity Load for Each Bus Line			150		pF
C_{I}	SCL, SDA Input Capacitance			5		pF
V_{IL}	Input Threshold Low				0.7	V
V_{IH}	Input Threshold High		1.4			V
I_{LKG}	Input Leakage Current	V = 0V and 5V	-1		1	uA
V_{OL}	Output Logic Low	$I_{\text{OL}} = 12\text{mA}$			0.36	V

5 Application

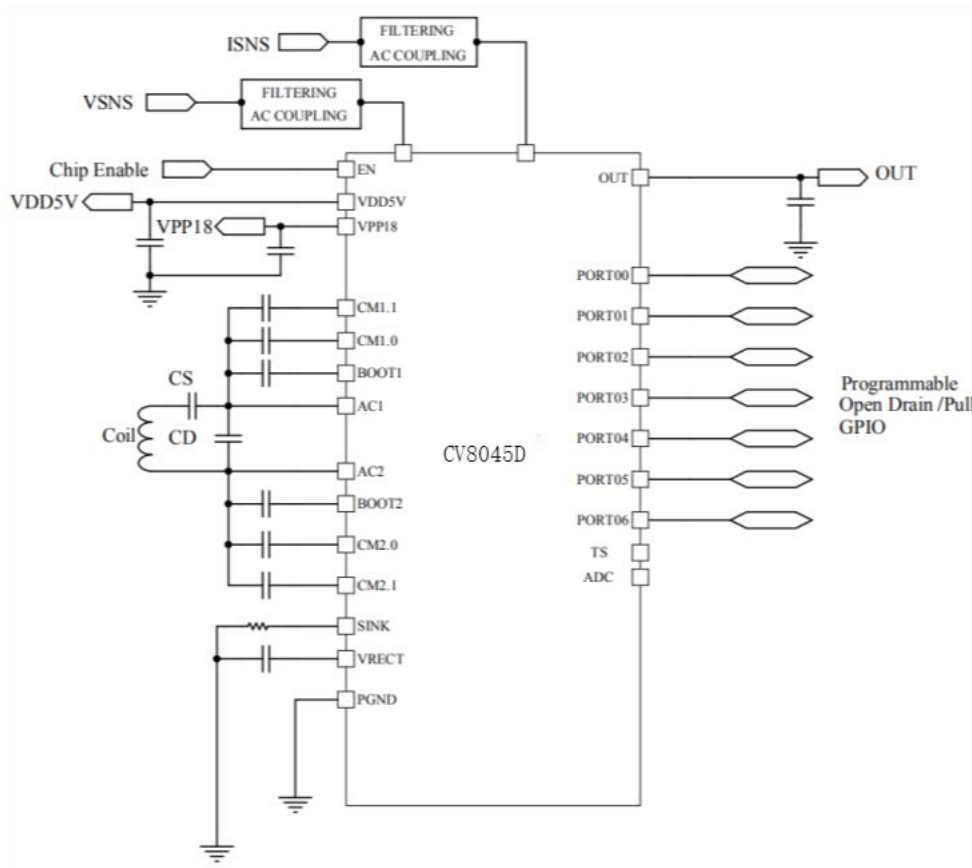


Figure 4

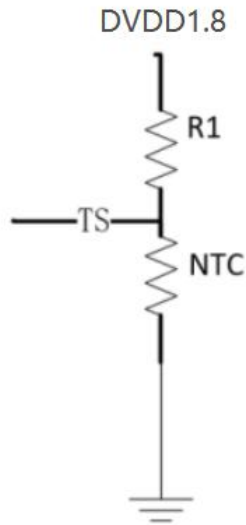
6 Setting and Configuration

6.1 SINK(Clamp1) pin

The CV8045D has embedded a programmable DC clamping to protect the device in the event of high voltage transients, which is a programmable current source, the dissipation capability are 40mA, 80mA, 120mA.

6.2 External Temperature sensing –TS

has a temperature sensor input, TS, which can be used to monitor an external temperature by using a thermistor.



The calculation formula as following: $V_{ts} = 1.8 \cdot NTC / (R1 + NTC)$

7 Typical Performance Characteristics

The following performance characteristics were taken using a CV90362 wireless power transmitter at $T_a = 25^\circ\text{C}$.

7.1 Efficiency 15W (12V@1.25A)

TX: CV90362, A11 TX Coil, $C_s = 400\text{nF}$, $d_z = 4\text{mm}$

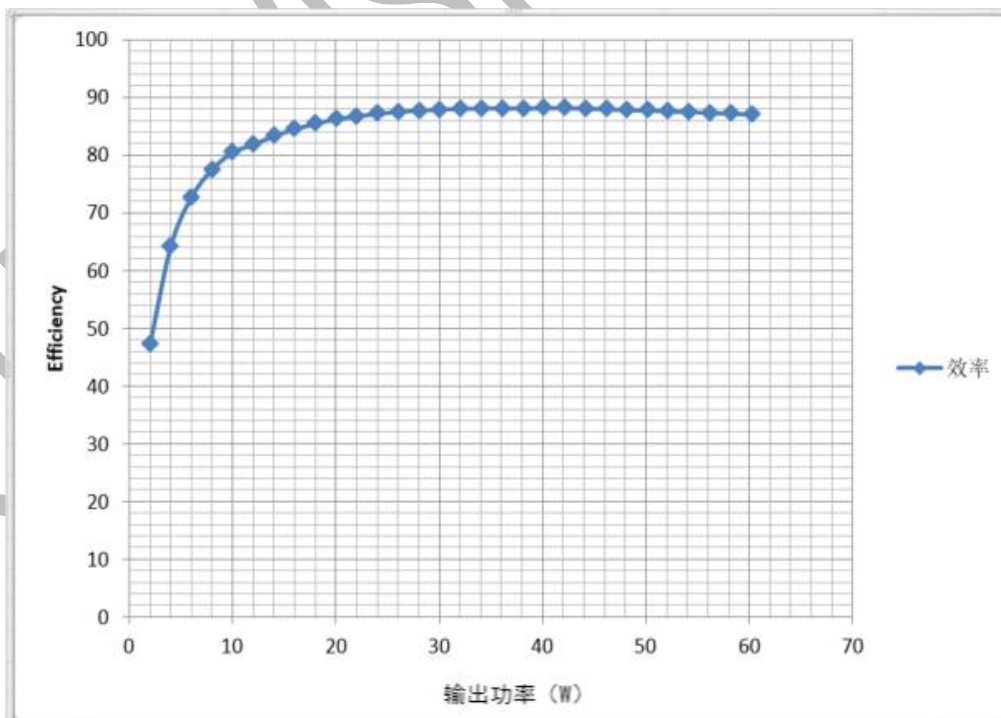


Figure 5. Efficiency vs. Output Power(W)

7.2 Transient Response

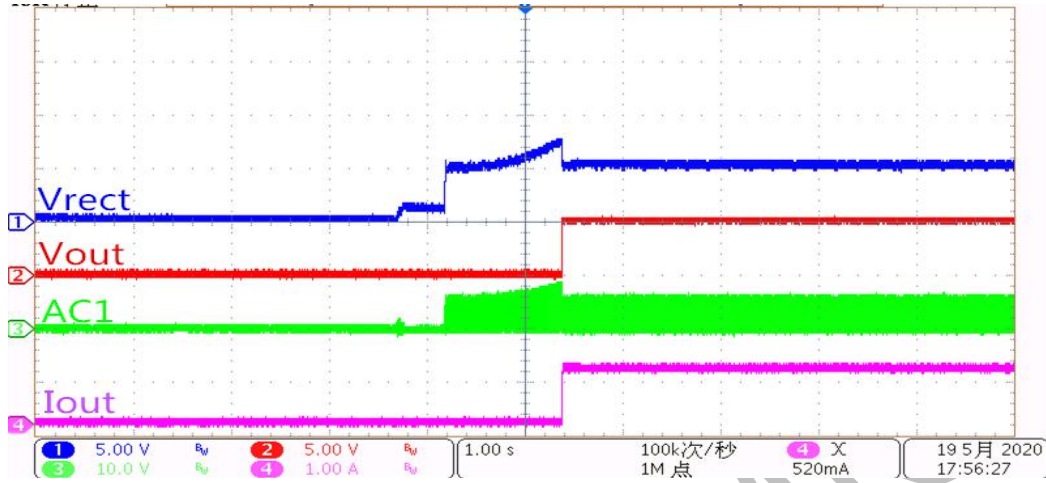


Figure 6. Transient Resp:CV8045D@Vout = 5V, Start up with 1A load

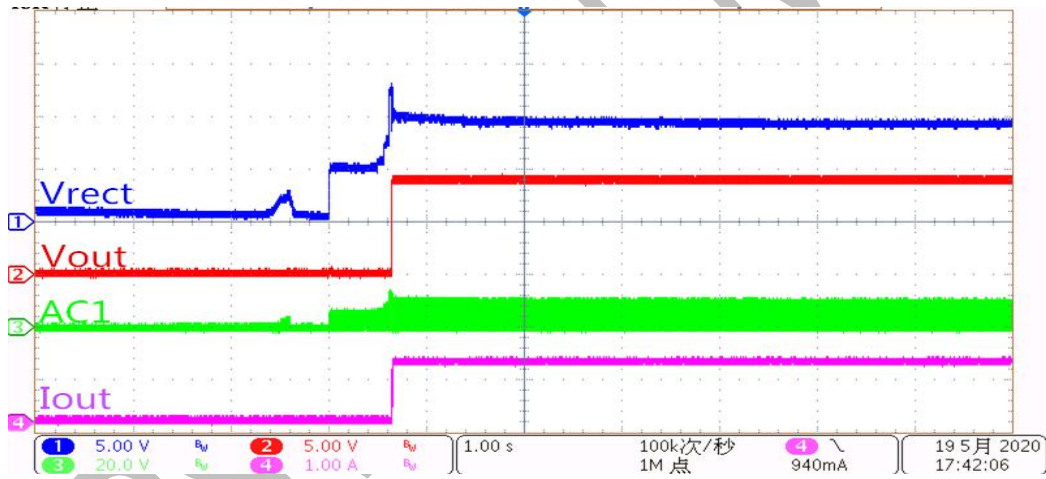


Figure 7. Transient Resp:CV8045D@Vout = 9V, Start up with 1.1A load

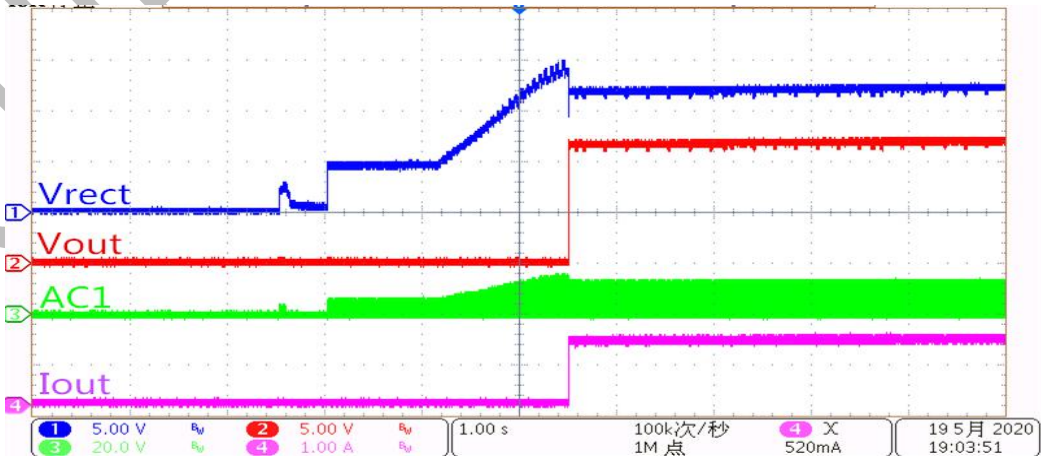


Figure 8. Transient Resp:CV8045D@Vout = 12V, Start up with 1.25A load

7.3 OVP protection

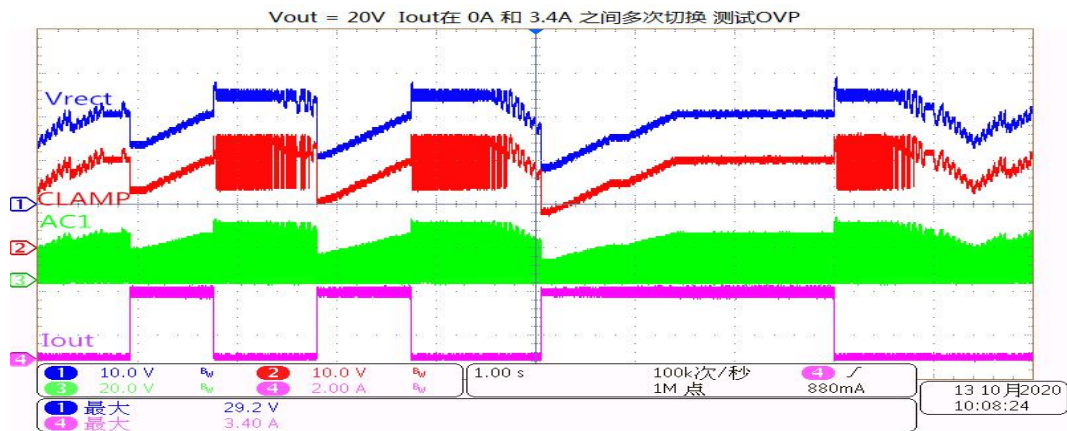


Figure 10. CV8045D@Vout = 20V, 3.4A to 0A load for testing OVP

7.4 OCP protection

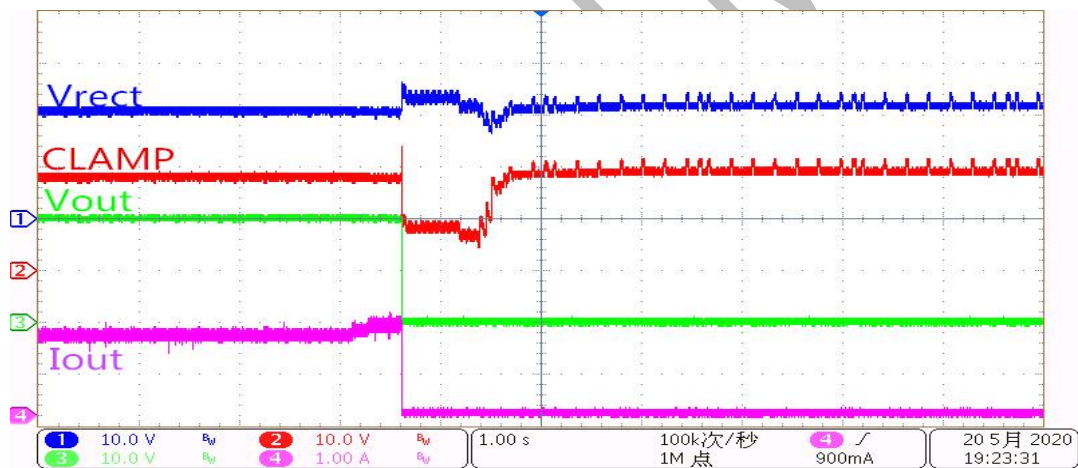


Figure 11. CV8045D@Vout = 20V, OCP @load = 1.7A

8 Application circuit

The circuit reference design is on page 19 of this document.

9 Bill of Materials (BOM)

Item	Comment	Designator	Libref	Quantity
1	0.1uF/100V	C1, C5, C6, C8	Cap	4
2	22nF/50V	C2, C10	Cap	2
3	10nF/50V	C3, C11	Cap	2
4	100nF/50V	C4, C12	Cap	2

5	3.3nF/100V	C7	Cap	1
6	0.1uF/10V	C13, C32	Cap	2
7	10uF/50V	C14, C15, C16, C17, C18, C21, C22, C23	Cap	8
8	0.1uF/50V	C19, C20	Cap	2
9	1uF/10V	C25,C26	Cap	2
10	10nF/25V	C27	Cap	1
11	5.6nF/100V	C28	Cap	1
12	1nF/25V	C29	Cap	1
13	6.8nF/10V	C30	CAP	1
14	4.7nF/10V	C31	CAP	1
15	IN4448	D1	DIODE	1
16	10K	R1, R2, R3, R10, R13,R16	RES	6
17	10k/1%	R4	RES	1
18	20K	R5	RES	1
19	47R 1/4W	R11, R12	RES	2
20	5.1K	R14	RES	1
21	200K	R15	RES	1
22	100K/B=4250	RNTC1	RES	1
23	CV8045D	U1	BGA	1
24	Coil	CL1	LOOP	1

10 Package info

The package outline drawings are appended at the end of this document.

11 Ordering Information

Orderable Part Number	Description and Package	MSL Rating	Shipping Packaging	Ambient Temperature
CV8045D	CV8045D Wireless Power Receiver for 15W & 5WTX Applications, 2.7 X 4.0 mm WLCSP-53	MSL1	Tape and reel	0°C to +85°C

